

DATE	DESCRIPTION	AMOUNT	BALANCE
1900	Jan 1		100.00
1901	Jan 1		100.00
1902	Jan 1		100.00
1903	Jan 1		100.00
1904	Jan 1		100.00
1905	Jan 1		100.00
1906	Jan 1		100.00
1907	Jan 1		100.00
1908	Jan 1		100.00
1909	Jan 1		100.00
1910	Jan 1		100.00
1911	Jan 1		100.00
1912	Jan 1		100.00
1913	Jan 1		100.00
1914	Jan 1		100.00
1915	Jan 1		100.00
1916	Jan 1		100.00
1917	Jan 1		100.00
1918	Jan 1		100.00
1919	Jan 1		100.00
1920	Jan 1		100.00
1921	Jan 1		100.00
1922	Jan 1		100.00
1923	Jan 1		100.00
1924	Jan 1		100.00
1925	Jan 1		100.00
1926	Jan 1		100.00
1927	Jan 1		100.00
1928	Jan 1		100.00
1929	Jan 1		100.00
1930	Jan 1		100.00
1931	Jan 1		100.00
1932	Jan 1		100.00
1933	Jan 1		100.00
1934	Jan 1		100.00
1935	Jan 1		100.00
1936	Jan 1		100.00
1937	Jan 1		100.00
1938	Jan 1		100.00
1939	Jan 1		100.00
1940	Jan 1		100.00
1941	Jan 1		100.00
1942	Jan 1		100.00
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1954	Jan 1		100.00
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1956	Jan 1		100.00
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1959	Jan 1		100.00
1960	Jan 1		100.00
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1965	Jan 1		100.00
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1968	Jan 1		100.00
1969	Jan 1		100.00
1970	Jan 1		100.00
1971	Jan 1		100.00
1972	Jan 1		100.00
1973	Jan 1		100.00
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1975	Jan 1		100.00
1976	Jan 1		100.00
1977	Jan 1		100.00
1978	Jan 1		100.00
1979	Jan 1		100.00
1980	Jan 1		100.00
1981	Jan 1		100.00
1982	Jan 1		100.00
1983	Jan 1		100.00
1984	Jan 1		100.00
1985	Jan 1		100.00
1986	Jan 1		100.00
1987	Jan 1		100.00
1988	Jan 1		100.00
1989	Jan 1		100.00
1990	Jan 1		100.00
1991	Jan 1		100.00
1992	Jan 1		100.00
1993	Jan 1		100.00

1. Field of the Invention

2. Description of Related Art

The input tuning circuit 41 in the UHF tuner 40 has a varactor diode 41a and first and second inductance elements 41b and 41c which are serially interconnected and connected in parallel with the varactor diode 41a. The cathode of the varactor diode 41a is grounded through a d.c. cut capacitor 41d and one end of the first inductance element 41b is also grounded. The junction of the two inductance elements 41b and 41c is connected to the

are converted into intermediate frequency signals by the mixer.

In receiving VHF band television signals, the VHF tuner 50 is activated and the UHF tuner 40 is inactivated; the VHF band television signals are selected and also converted into intermediate frequency signals by the VHF tuner 50.

It is well known that generally, the Q factor and bandwidth B of a tuning circuit that varies the tuning frequency by means of a varactor diode are expressed by equations $Q = R/\omega L$ and $B = L \times \omega^2/2\pi R$, respectively. Therefore, even when the tuning impedance R (resistance) is constant, the bandwidth B broadens as the frequency increases. In addition, since the impedance of the coupling capacitor 42 in the above conventional tuner circuit configuration varies depending on the frequency, the tuning circuit's loss resistance R including the converted impedance is smaller as the tuning frequency is higher; as a consequence, variation in the bandwidth B is considerable and as shown in Fig. 5, the tuning characteristic curve which represents selectivity is steep (i.e. narrow bandwidth) and insertion loss S is large in UHF low band frequencies while the insertion loss is small and the tuning characteristic curve is gradual (broad bandwidth) in UHF high band frequencies.

Consequently, when low band UHF television signals are received, NF (noise figure) deteriorates; when high

band UHF television signals are received, interference by television signals from a neighboring channel occurs.

SUMMARY OF THE INVENTION

In view of the above circumstances, the present invention provides a television tuner which prevents interference by a neighboring channel, particularly in receiving high band television signals, by maintaining the tuning circuit bandwidth virtually constant throughout the UHF low to high band range, and thereby improves the NF in receiving low band television signals, eliminating a difference in NF between channels.

In order to achieve the above object, the tuner based on the present invention comprises the following: an input terminal through which UHF band or VHF band television signals are inputted; a UHF tuner which receives the UHF band television signals; and a VHF tuner which is provided together with the UHF tuner and receives at least the VHF band television signals, where the UHF tuner is composed of a UHF tuning circuit having a varactor diode and first and second inductance elements serially interconnected and connected in parallel with the varactor diode, where the varactor diode varies a tuning frequency within a prescribed frequency range in the UHF band; and impedance means serially inserted between the input terminal and the junction of the two inductance elements, where the impedance by the impedance means increases as a frequency

increases in the prescribed frequency range.

In addition, the impedance means consists of a series resonance circuit composed of a third inductance element and a capacitance element, where a resonance frequency of the series resonance circuit is set to below the minimum frequency in the prescribed frequency range.

A diode is serially inserted in the impedance means and the diode is turned on when UHF band television signals are received, while it is turned off when VHF band television signals are received.

Furthermore, one end of the first inductance element is grounded and a resonance frequency of the series circuit composed of the diode in the off state, the series resonance circuit and the first inductance element is set to within the UHF band.

Furthermore, a band switching circuit is provided to generate UHF selection voltage and VHF selection voltage which activate or inactivate the UHF tuner and the VHF tuner, respectively, and the UHF selection voltage is applied to the anode of the diode and the VHF selection voltage is applied to its cathode.

BRIEF DESCRIPTION OF THE DRAWINGS

A preferred embodiment of the present invention will be described based on the followings, wherein:

Fig. 1 is a circuit diagram showing the configuration of a television tuner according to the

present invention;

Fig. 2 is a graph showing the UHF band tuning characteristic of the television tuner according to the present invention;

Fig. 3 is a graph showing the VHF band tuning characteristic of the television tuner according to the present invention;

Fig. 4 is a circuit diagram showing the configuration of a conventional television tuner; and

Fig. 5 is a graph showing the UHF band tuning characteristic of the conventional television tuner.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

Fig. 1 shows the configuration of a television tuner according to the present invention. UHF band television signals and VHF band television signals are inputted through an input terminal 1 from an antenna not shown. The input terminal 1 is connected through a highpass filter 2 to a UHF tuner 10 and a VHF tuner 20. The UHF tuner 10 receives signals of approximately 470 MHz to 800 MHz while the VHF tuner 20 receives VHF band signals and some of UHF band low band signals. The highpass filter 2 allows VHF band or higher frequency signals to pass through it and attenuates intermediate or lower frequency signals; particularly 27 MHz band signals as used for citizen band communication equipment are attenuated by a series resonance circuit 2a.

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A UHF input tuning circuit 11 of the UHF tuner 10 has a varactor diode 11a and first and second inductance elements 11b and 11c serially interconnected and substantially connected in parallel with the varactor diode. The cathode of the varactor diode 11a is grounded through a d.c. cut capacitor 51d in accordance with high frequency grounding practices and one end of the first inductance element 11b is also grounded. The varactor diode 11a is connected in parallel with another inductance element 11e, which is intended to compensate for an undesired decrease in the inductance value of the first and second inductance elements 11b and 11c.

Between the junction T of the two inductance elements 11b and 11c, and the highpass filter 2, an impedance means 12, which consists of a series resonance circuit with a third inductance element 12a and a capacitance element 12b as a coupling capacitor, and a diode 13 are interconnected in a series. The resonance frequency of the impedance means 12 is set to below the UHF band low band frequency (470 MHz). As a result, the impedance in the direction from the junction T to the input terminal 1 is converted into a high impedance according to the ratio of the two inductance elements 11b and 11c (a logical explanation of this conversion is omitted here). This is equivalent to the resistance component in the converted impedance being connected in parallel with the UHF input tuning circuit 11.

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The orientation of the diode 13 is not limited; in this case, the anode is on the side of the highpass filter 2 and the cathode is on the side of the impedance means 12. VHF selection voltage V_b from a band switching circuit 3 is applied to the cathode of the diode 13. When the diode 13 turns off, it has a capacitance of 0.5 pF to 1.0 pF, so the diode 13 in the off state, the impedance means 12 and the first inductance element 11b make up a series resonance circuit. The resonance frequency of the circuit is set so as to be within the UHF band.

The tuning frequency of the UHF input tuning circuit 11 varies within a prescribed frequency range, for example, from approximately 470 MHz to 800 MHz depending on the tuning voltage V_t applied to the cathode of the varactor diode 11a. A UHF high frequency amplifier 14 is connected next to the UHF input tuning circuit 11. The UHF high frequency amplifier 14 uses an FET 14a and a peaking coil 14b is connected between the first gate as its input terminal and the ground. The peaking coil 14b compensates for a gain drop of the FET 14a in the UHF band low band (around 470 MHz). UHF selection voltage U_b from the band switching circuit 3 is applied to the first gate of the FET 14a through the peaking coil 14b. A UHF mixer or the like is located after the UHF high frequency amplifier 14, though it is not shown in the figure.

The VHF tuner 20 comprises: a trap circuit 21 connected to the highpass filter; a VHF input tuning

bandwidth are uniform throughout the whole (low to high) band range.

When the VHF tuner 20 is active, UHF band signals are attenuated by the diode 13, the impedance means 12 and the first inductance element 11b, as shown in Fig. 3. This minimizes interference by UHF band television signals when the VHF tuner 20 receives high band television signals.

As discussed so far, the UHF tuner comprises: a UHF tuning circuit which has a varactor diode and a first and a second inductance element serially interconnected and connected in parallel with the varactor diode where the tuning frequency varies within a prescribed frequency range in the UHF band; and an impedance means which is serially inserted between the input terminal and the junction of the two inductance elements, where the impedance by the impedance means increases as the frequency increases within the prescribed frequency range in the UHF band. This reduces insertion loss in the UHF low band and suppresses a bandwidth increase in the UHF high band. So, uniformity in insertion loss and bandwidth is ensured throughout the whole (low to high) band range, which improves NF in the low band and reduces interference by a neighboring channel in the high band.

In addition, the impedance means consists of a series resonance circuit composed of a third inductance element and a capacitance element, where the resonance

UHF tuner and turned off to activate the VHF tuner.

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